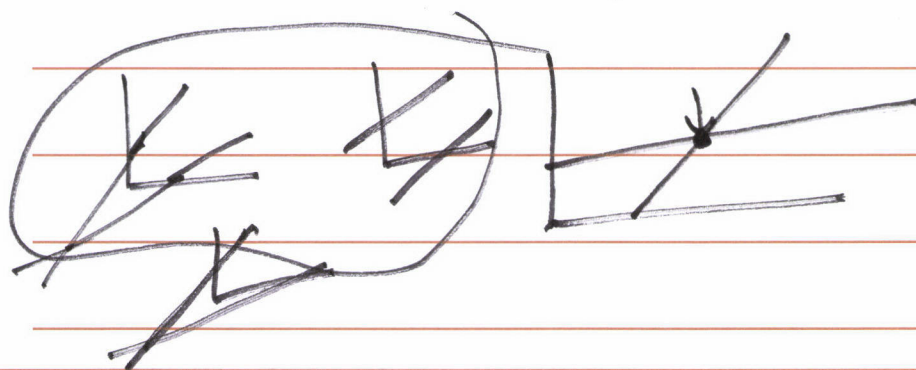


Power Control

Independent channels, allocating power levels to satisfy SINR threshold in fixed rate radios.



Corrected expressions for optimal power allocation

$$P_1^* = \frac{N \left(\frac{1}{g_{21}} + \frac{\theta}{g_{22}} \right)}{\left(\frac{g_{21}}{g_{21}\theta} - \frac{g_{12}}{g_{22}} \right)}$$

$$P_2^* = \frac{N \left(\frac{1}{g_{12}} + \frac{\theta}{g_{11}} \right)}{\left(\frac{g_{22}}{g_{12}\theta} - \frac{g_{21}}{g_{11}} \right)}$$

Generalize to M links

$$\frac{P_i g_{ii}}{\sum_{\substack{j=1 \\ j \neq i}}^M P_j g_{ji}} + N \geq \theta \quad \forall i$$

$$P_i g_{ii} \geq \sum_{\substack{j=1 \\ j \neq i}}^M \theta g_{ji} P_j + N\theta$$

$$P_i g_{ii} - \sum_{j=1}^M \theta g_{ji} P_j \geq N\theta \quad \forall i = 1:M$$

At the minimum power vector solution (if it exists), all of these become equalities.

$$\begin{array}{c}
 \vec{A} \vec{x} = \vec{b} \\
 \uparrow \quad \uparrow \quad \uparrow \\
 M \times M \quad M \times 1 \quad M \times 1
 \end{array}$$

jth

\vec{P}
 \downarrow
 x
 \downarrow

$$A = \begin{bmatrix}
 \delta_{11} & & & & \\
 & \delta_{22} & & & \\
 & & \delta_{ii} & -\delta_{ji} & \dots \\
 & & & \dots & \\
 & & & & \dots & \delta_{mm}
 \end{bmatrix}
 \begin{bmatrix}
 P_1 \\
 P_2 \\
 \vdots \\
 P_i \\
 P_j \\
 \vdots \\
 P_m
 \end{bmatrix}$$

\vec{b}

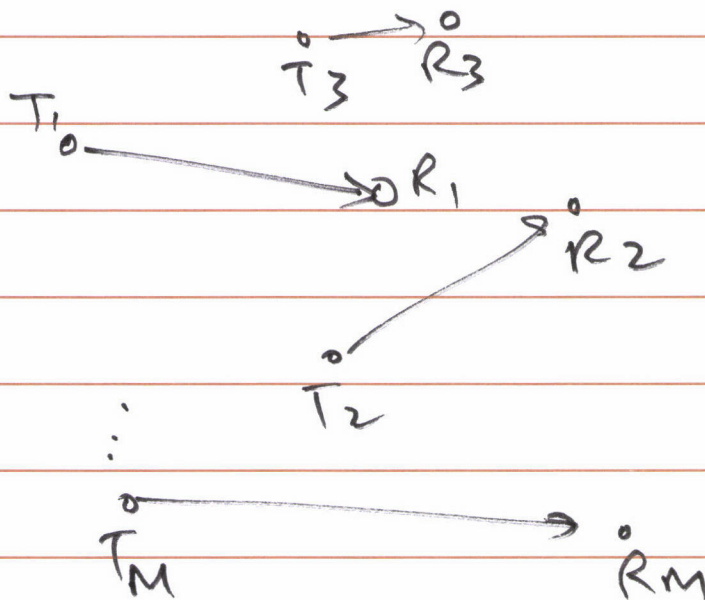
$$= \begin{bmatrix}
 N0 \\
 \vdots \\
 N0
 \end{bmatrix}$$

$$A \vec{P} = \vec{b}$$

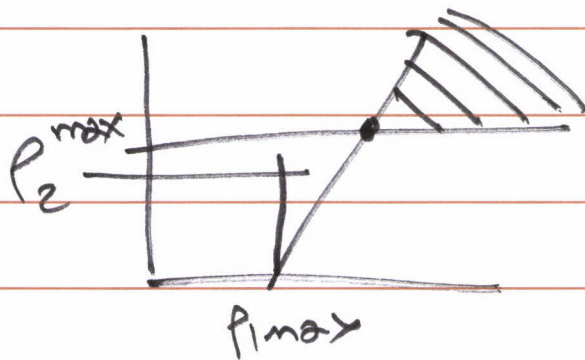
$$A^{-1} A \vec{P} = A^{-1} \vec{b}$$

$$\boxed{\vec{P} = A^{-1} \vec{b}}$$

- A should be invertible
 i.e. have non-zero determinant
- $\vec{P}^* \geq 0$
 elementwise

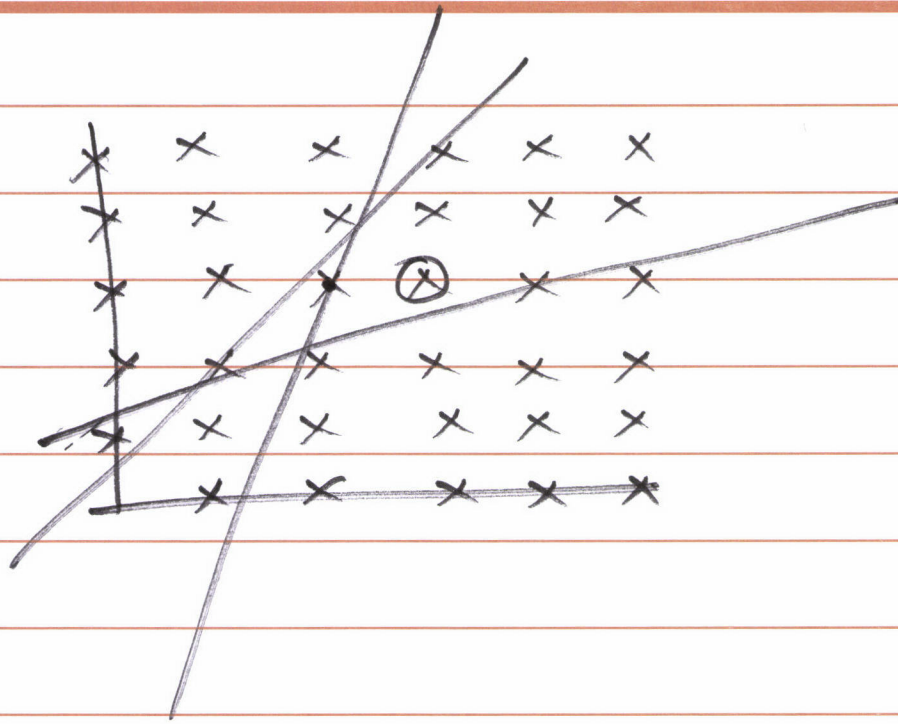


Additional conditions typical in real radios :



- $P_i^* \leq P_i^{\max}$

- P_i may only be allowed to take on discrete values.



Two issues to consider

- Distributed implementation
- What to do w/ infeasible situations?

There are 2^M possible configurations of M links.

In general finding the maximum sized configuration is computationally hard (NP-hard). Typically some kind of a greedy heuristic such as "turn off most-interfering transmitter" is used.

A related problem is finding the minimum # of time slots/frequencies to schedule a given set of M links.

Foschini - Miljanic Algorithm for Iterative, Distributed Power Control

$$P_i^{t+1} = P_i^t \cdot \frac{\Theta}{\text{SINR}_i(t)}$$

$$\text{if } \text{SINR}_i(t) = \frac{P_i^t \cdot g_{ii}}{\sum_{j \neq i} P_j^t \cdot g_{ji} + N} > \Theta$$

the ^(link) node i will reduce its transmit power at time $t+1$.

$$\text{if } \text{SINR}_i(t) < \Theta$$

the transmitting node i will increase its power at time $t+1$.

Can prove that this algorithm
converges to the global
optimum power vector
 \vec{p}^*

as $t \rightarrow \infty$

In practice 2 to 3 iterations
often suffice to get near-optimal,

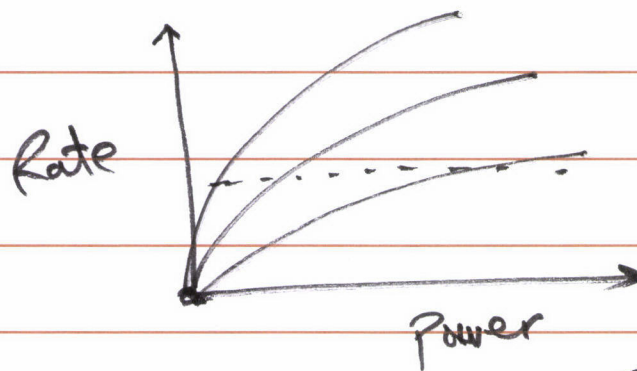
feasible power levels.

Power Allocation for
 Maximizing sum rate over
 parallel channels (assuming a
 rate-adaptive radio model).

$$\max \sum_{i=1}^M \log(1 + \text{SNR}_i)$$

$$\text{s.t.} \quad \sum_{i=1}^M P_i \leq P_{\text{TOT.}}$$

$$P_i \geq 0 \quad \forall i$$

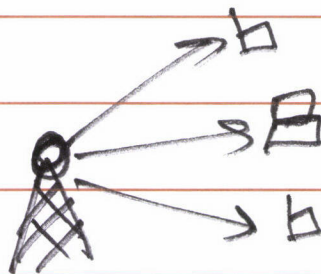


(no interference)

$$\text{SNR}_i = P_i \cdot \left(\frac{g_i}{N_i} \right)$$

gain to noise ratio.

Multimuser
 e.g. OFDM
 in WiMax



This is a convex optimization problem.

Solving this problem requires the use of Lagrange multipliers to handle the constraints.

Can simplify the problem to

$$\begin{aligned} \max \quad & \sum_{i=1}^M \log(1 + P_i \frac{g_i}{N_i}) \\ \text{s.t.} \quad & \sum_{i=1}^M P_i = P_{\text{TOT}} \end{aligned}$$

$$\max f(\bar{x}) \quad \text{s.t.} \quad g_j(\bar{x}) \leq c_j$$

$$\mathcal{L}(\bar{x}, \bar{\lambda}) = f(\bar{x}) - \sum_j \lambda_j (g_j(\bar{x}) - c_j)$$

find \bar{x}^* , $\bar{\lambda}^*$ that satisfies

KKT conditions:

$$\lambda_j^* \geq 0$$

$$g_j(\bar{x}^*) \leq c_j \quad \text{complementary slackness}$$

$$\frac{\partial \mathcal{L}}{\partial x_i} \Big|_{\bar{x}^*, \bar{\lambda}^*} = 0$$

$$\lambda_j^* (g_j(\bar{x}^*) - c_j) = 0$$

$$-P_i \leq 0 \Leftrightarrow P_i \geq 0 \quad \forall i$$

$$\max \sum \log(1 + P_i g_i / N_i) \quad \text{s.t.} \quad \sum P_i \leq P_{\text{tot}}$$

$$\mathcal{L}(P, \lambda) = \sum_{i=1}^M \log(1 + P_i g_i / N_i) - \lambda (\sum P_i - P_{\text{tot}})$$

$$\lambda \geq 0 \quad \sum P_i \leq P_{\text{tot}} \quad \nu_i \geq 0 \quad + \sum \nu_i P_i$$

$$\frac{\partial \mathcal{L}}{\partial P_i} = \frac{g_i / N_i}{(1 + P_i^* g_i / N_i)} - \lambda^* + \nu_i^* = 0$$

$$\lambda^* (\sum P_i^* - P_{\text{tot}}) = 0 \quad \nu_i^* \cdot P_i^* = 0$$

$$\sum P_i^* \leq P_{\text{tot}} \\ P_i^* \geq 0$$

complementary slackness conditions

for any given channel, either $v_i^* = 0$
or $P_i^* = 0$.

focus on a channel i that gets
non-zero power, its $v_i^* = 0$

$$\frac{g_i/N_i}{(1 + P_i^* g_i/N_i)} - \lambda^* + v_i^* = 0$$

$$\lambda^* = g_i/N_i / (1 + P_i^* g_i/N_i)$$

$$P_i^* = \left(\frac{g_i/N_i}{\lambda^*} - 1 \right) \cdot \frac{N_i}{g_i}$$

$$P_i^* = \left(\frac{1}{\lambda^*} - \frac{N_i}{g_i} \right)$$

$$P_i^* = \left(\frac{1}{\lambda^*} - \frac{N_i}{g_i} \right)^+ \quad \text{--- ①}$$

$$\sum_{i=1}^M P_i^* = P_{TOT} \quad \text{--- ②}$$

$$\sum_{i=1}^M \left(\frac{1}{\lambda^*} - \frac{N_i}{g_i} \right)^+ = P_{TOT}$$

an implicit equation that defines what λ^* should be.

This solution is called **waterfilling**

